#### LOGICAL DESIGN OF A REDUNDANT BINARY ADDER

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## Abstract

This paper investigates the logical design of a redundant binary adder with two input digits and one output digit, all in the digit set  $\{\overline{1}, 0, 1\}$ . Redundant binary arithmetic structures in which all digit sets are  $\{\overline{1}, 0, 1\}$  were first discussed by Avizienis in 1961. Borovec studied the logical design of a class of such binary adders and subtracters in 1968. At that time, a variation of the adder/subtracter was overlooked. This paper studies the logical design of this variation. The sum digit is still a function only of the digits in three adjacent digital positions of the operands. "Coupled don't cares" are encountered, but have not introduced too much difficulty. The nine distinct formats (under permutation and negation) of representing three values with two bits given by Robertson are used. The simplest adder/ subtracter designs from this variation are less complex than the simplest designs previously known.

## Background

This paper investigates the logical design of a redundant binary adder with two inputs and one output, all in the digit set  $\{\bar{1}, 0, 1\}$ .

The available hardware technologies have certainly made the implementation of redundant arithmetic algorithms practical. The redundant binary adder is a basic building unit in redundant arithmetic structures whether they perform fast digit parallel operations without propagating transfers or most-significant-digit-first digit serial operations 1,2. The redundant binary adder is heavily used in Goyal's LSI implementable arithmetic unit. Goyal also showed that radix-2<sup>k</sup> operations can be easily implemented with redundant binary adders. Although LSI technologies have made logic design minimization less urgent, the redundant binary adder which may be used hundred of times even in a modest

arithmetic unit deserves to be studied in detail. With the advent of digit pipelining algorithms which inevitably require a redundant number representation, the authors are convinced that the redundant binary adder will be a fundamental structure in the future.

A redundant arithmetic structure in which all digit sets are {1, 0, 1} was first discussed by Avizienis. 3 He showed that the requirements of a totally parallel operation and a unique representation of zero dictate a three-level structure for the signed-digit binary adder/subtracter. The result digit depends only on digits in three adjacent digital positions of the operands. Borovec investigated the logical design of this structure. 4 One of his two approaches (Figure 1) is to split the sum of the operand digits in the i th position, 4 and ki, into ai-l and mi such that

% i + ki = 2 · ai - l + mi
and input these to adder/subtracter
structures which operate on one redundant
digit and one nonredundant digit.
(An asterisk beside a symbol denotes
a redundant binary digit. (i-l) is the
position of the left neighboring unit.)
His second approach consists of combining
the two lower levels so that the format of
ai need not be considered (Figure 2). This
paper studies the structures resulting from
having different outputs from the lowest
level. The sum of the pair of operand
digits in the ith position is expressed as
(2·mi-l + ai) as illustrated in Figure 3.

## The Adder Structure

Since it is not trivial to negate a number expressed in the normalized redundant binary digit set, the operands digits  ${\overset{*}{\iota}}_i^*$  and  $k_i^*$  are restricted to the symmetric digit set  $\{\overline{1},\ 0,\ 1\}$ . The structure shown

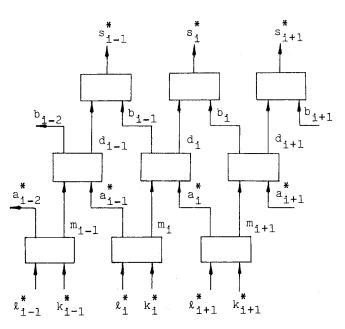


Figure 1 Redundant binary adder/
subtracter structure with
three separate levels

Figure 2 Combining the two lower levels

in Figure 3 can again be built either in three separate levels or with the two lower levels combined. However, the arrangement with three separate levels is not very useful in this case because the valid groups of digit sets for  $m_i$  and  $a_i^*$  are not practical. The sum of  $\ell_i^*$  and  $k_i^*$  equals  $(2 \cdot m_{i-1} + a_i^*)$ . Hence, the digit sets for  $m_{i-1}$  and  $a_i^*$  must be from one of the following combinations:

1.  $m_{i-1} \in \{\overline{1}, 0\}, a_i^* \in \{0, 1, 2\},$ 

2.  $m_{i-1} \in \{0, 1\}, a_i^* \in \{\overline{2}, \overline{1}, 0\}, or$ 

3.  $2 \cdot m_{i-1} \in \{\overline{1}, 1\}, a_i^* \in \{\overline{1}, 0, 1\}.$ 

For the design with three separate levels, one would like to use the same digit set for the redundant transfer digit, the two operand digits and the sum digit

(a\*\* (\*\* k\*\* and s\*\* respectively) and to

 $(a_i^*, k_i^*, k_i^*, and s_i^* respectively)$  and to have  $m_{i-1}$  as a conventional binary digit.

Failing to satisfy these, the threeseparate-levels arrangement needs no further attention.

Combining the two lower levels does result in relatively simple logic designs (Figure 4). For the symmetric adder, the algebraic relationships are

1.  $d_i + b_i = s_i$  for the upper block, and

2. 
$$\ell_{i}^{*} + k_{i}^{*} + m_{i} = 2 \cdot m_{i-1} + 2 \cdot b_{i-1} + d_{i}$$
for the lower block.

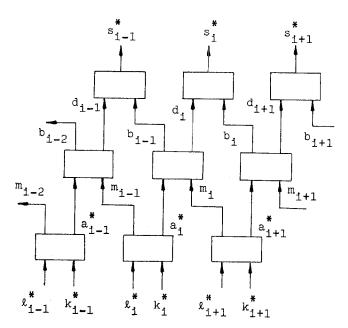


Figure 3 Variation of the three level structure

Inspite of the above equation,  $m_{i-1}$  can be made independent of  $m_i$ . The final \*\* sum digit  $s_i^*$  is still a function only of the digits in three adjacent digital positions of the operands. The chosen sets for  $m_i$ ,  $b_i$ ,  $d_i$  are  $\{0, 1\}$ ,  $\{1, 0\}$ ,  $\{0, 1\}$  respectively. Before actually designing the logic circuit, one has to fix the representation of the redundant binary digit. Robertson<sup>5</sup> has shown that there exist only nine distinct ways, under permutation and negation, of representing three values with two bits. The nine formats are shown in Table 1. Since it is sometimes necessary to feed the output from an adder back as an operand, the sum digit \*s\_i^\* should have the same format as the operand digits  $l_i^*$  and  $k_i^*$ .

## Details

Some rather interesting details are encountered in the logic designing of this adder. Whenever  $(m_i + \ell_i^* + k_i^*)$  is algebraically 0 or 1,  $(-b_{i-1})$  and  $m_{i-1}$  can be both 0 or both 1. This results in  $b_{i-1}$  and  $m_{i-1}$  having "coupled don't cares" in the truth tables. As an example, Table 2

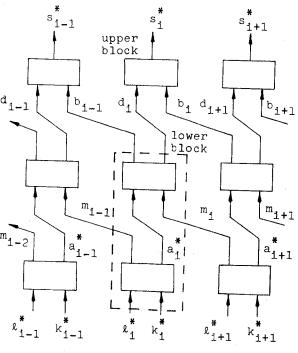


Figure 4 Combining the two lower levels of Figure 3

shows the truth table for the lower block of the adder using format 2 from Table 1. \* \* \* and s \* are represented by pairs of bits  $(\lambda_i, \ell_i)$ ,  $(\kappa_i, k_i)$  and  $(\sigma_i, s_i)$  respectively. The horizontal pairs of c's

s <sub>i</sub>			PN distinct				formats				
σ <sub>i</sub>	s <sub>i</sub>	1	2	3	4	5	6	7	8	9	
0	0	0	0	0	0	·d	0	0	1	ī	
С	1	1	1	1	1	1	1	1	1	1	
1	0	1	d	0	ī	0	1	ī	0	0	
1	1	0	ī	ī	d	ī	ī	ī	1	ī	

Table 1 Nine PN (permutation-negation)
distinct formats of representing
a redundant binary digit with
two bits

							T-1-		
	mi	λ <sub>i</sub>	li	κ <sub>i</sub>	k	value	m <sub>i-1</sub>	b <sub>i-</sub>	d <sub>i</sub>
0 1 2 3	0 0 0 0	0 0 0	0 0 0	0 0 1 1	0 1 0 1	0 1 <u>d</u> 1	c c(1) d 0	c c(1) d 1	0 1 d
4 5 6 7	0 0 0	0 0 0 0	1 1 1	0 0 1	0 1 0 1	1 2 d 0	c(1) 1 d		1 0 d 0
8 9 10 11	0 0 0	1 1 1	0 0 0 0	0 0 1 1	0 1 0 1	d d d	d d d	d d d	d d d d
12 13 14 15	0 0 0	1 1 1	1 1 1	0 0 1	0 1 0 1	1 0 <u>d</u> 2	0 c d	1 c d	1 0 d
16 17 18 19	1 1 1	0 0 0	0 0 0	0 0 1	0 1 0 1	1 2 d 0	c 1 d c(0)	c 0 d c(0)	1 0 d
20 21 22 23	1 1 1		1	0 0 1 1	0 1 0 1	2 3 d 1	1 1 d c	0 0 d	0 1 d 1
24 25 26 27	1 1 1 1	1 1	0	0 0 1	0 1 0 1	d d d	d d d	d d d	đ đ đ
28 29 30 31	1	1 1 1 1	1 (	0 1 1	0 1 0	0 1 <u>d</u> 1	c(0) c d 0	c(0) c d 1	0 1 d

Table 2 Truth table for the lower block of format 2

across  $\mathbf{m_{i-l}}$  and  $\mathbf{b_{i-l}}$  are the "coupled don't cares". A horizontal pair of c's can be both 0 or both 1.

The complexity introduced by the "coupled don't cares" is greatly reduced by the important constraint that the transfers  $m_{i-1}$  and  $b_{i-1}$  be non-propagating. This means that  $\mathbf{m}_{\text{i-l}}$  must be independent of  $\mathbf{m}_{\text{i}}$ and fixes the values of most of the "coupled don't cares". In Table 2, the "coupled don't cares" for minterms 1, 4, and 28 must have the values 1, 1, 0, and 0 respectively because the upper 16 function values of m<sub>i-l</sub> must match the lower 16 function values of  $m_{i-1}$ . Figure 5 shows the Karnaugh maps for  $\mathbf{m_{i-1}}$  from Table 2. Besides the coupled relationship between m<sub>i-l</sub> and b<sub>i-l</sub>, the "coupled don't cares" are also coupled across the maps corresponding to  $m_i = 0$  and  $m_i = 1$  as indicated by the dotted lines in Figure 5. For this particular format, there remain only 8 possible assignments of the values of  $m_{i-1}$ . The simplest design for this format is given in Table 3.

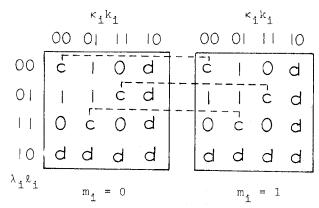


Figure 5 Karnaugh maps for  $m_{i-1}$  for the adder with format 2

The other formats for the redundant binary digit are similarly studied. The criterion used for choosing a design for each format is that the boolean functions yield a minimal total number of literals when they are expressed as sums of products or products of sums.

If the chosen digit sets for m<sub>i</sub>, b<sub>i</sub> and d<sub>i</sub> are negated, the resulting combination of digit sets is still valid. However, there is no need to investigate the logic designs of the negated digit sets

because they are similar in complexity to the designs of the original digit sets. For each of the nine formats, the truth table for the negated digit sets can be obtained from the truth table of the chosen digit sets by changing  $\textbf{m}_i$  to  $\overline{\textbf{m}}_i$  and switching the output functions  $\textbf{b}_{i-1}$  and  $\textbf{m}_{i-1}$ .

Table 4 shows the designs for the symmetric subtracter. For the symmetric subtracter, the algebraic relationship for the lower block of Figure 4 is

 $m_i$  +  $l_i^*$  -  $k_i^*$  =  $2 \cdot m_{i-1}$  +  $2 \cdot b_{i-1}$  +  $d_i$ . The upper block is the same as that of the adder. The nine formats are again used to yield the designs shown in Table 4.

The simplest designs both for the adder and the subtracter are those corresponding to formats 2, 4 and 5. These designs are simpler than the best designs given in reference 4 (the appendix).

1. 
$$d_{1} = m_{1} \bigoplus_{i} \lambda_{i} \bigoplus_{i} \lambda_{i}$$

Table 3(a) Boolean equations for symmetric adders with formats 1 to 4

$$\begin{bmatrix} 5. & d_{1} = m_{1} \bigoplus \ell_{1} \bigoplus k_{1} \\ m_{1-1} = \overline{\lambda_{1}} \overline{k_{1}} \vee \overline{\kappa_{1}} \\ b_{1-1} = \lambda_{1} k_{1} (\overline{m_{1}} \vee \ell_{1}) \vee \overline{m_{1}} \ell_{1} \overline{k_{1}} \end{bmatrix}$$

$$\sigma_{1} = \overline{d_{1}} \vee b_{1}$$

$$\sigma_{1} = \overline{d_{1}} \vee b_{1}$$

$$\delta. & d_{1} = m_{1} \bigoplus \overline{\lambda_{1}} \overline{k_{1}} \bigoplus \overline{\kappa_{1}} \overline{k_{1}} \\ m_{1-1} = \overline{\lambda_{1}} \overline{k_{1}} \vee \overline{\ell_{1}} \overline{\kappa_{1}} \vee \kappa_{1} \overline{k_{1}} \\ b_{1-1} = \overline{\lambda_{1}} \overline{\ell_{1}} (\overline{m_{1}} \vee \overline{\kappa_{1}} \overline{k_{1}}) \vee \overline{m_{1}} \overline{\kappa_{1}} \overline{k_{1}} \\ m_{1-1} = \overline{\lambda_{1}} \overline{\ell_{1}} (\overline{m_{1}} \vee \overline{\kappa_{1}} \overline{k_{1}}) \vee \overline{m_{1}} \overline{\kappa_{1}} \overline{k_{1}}$$

$$\sigma_{1} = b_{1} \overline{d_{1}} \\ s_{1} = d_{1} \bigoplus b_{1}$$

$$7. & d_{1} = m_{1} \bigoplus \overline{\lambda_{1}} \overline{k_{1}} \\ b_{1-1} = \overline{\lambda_{1}} \overline{k_{1}} (\overline{m_{1}} \vee \overline{\kappa_{1}} \overline{k_{1}}) \vee \lambda_{1} \kappa_{1} \vee \overline{m_{1}} \overline{\kappa_{1}} \overline{k_{1}}$$

$$\sigma_{1} = \overline{d_{1}} \oplus b_{1}$$

$$7. & d_{1} = m_{1} \bigoplus \overline{\lambda_{1}} \overline{k_{1}} \\ b_{1-1} = \overline{\lambda_{1}} \overline{k_{1}} (\overline{m_{1}} \vee \overline{\kappa_{1}} \overline{k_{1}}) \vee \lambda_{1} \kappa_{1} \vee \overline{m_{1}} \overline{\kappa_{1}} \overline{k_{1}} \overline{k_{1}}$$

$$\sigma_{1} = \overline{d_{1}} \oplus b_{1}$$

$$8. & d_{1} = \lambda_{1} \overline{\ell_{1}} \bigoplus \kappa_{1} \overline{k_{1}} \bigoplus m_{1} \\ m_{1-1} = \overline{\lambda_{1}} \vee \overline{\kappa_{1}} \vee \overline{\kappa_{1}} \vee \overline{k_{1}} \vee \overline{k_{1}} \vee \overline{k_{1}} \vee \overline{k_{1}} \vee \overline{k_{1}} )$$

$$\sigma_{1} = \overline{d_{1}} \vee b_{1}$$

Table 3(b) Boolean equations for symmetric adders with formats 5 to 9

⊕ denotes "exclusive or";
∨ denotes "or";
juxtaposition or · denotes "and".

 $\sigma_{i} = \overline{d}_{i} \vee b_{i}$   $s_{i} = d_{i} \oplus b_{i}$ 

1. 
$$d_{1} = m_{1} \oplus \lambda_{1} \oplus \lambda_{1} \oplus \kappa_{1} \oplus \kappa_{1} \oplus \kappa_{1}$$
 $m_{1-1} = (\lambda_{1} \vee \kappa_{1})(\overline{\lambda_{1}} \vee \overline{\kappa_{1}})$ 
 $b_{1-1} = \overline{m_{1}}(\lambda_{1} \oplus k_{1}) \vee \lambda_{1} \overline{x_{1}} \overline{\kappa_{1}} k_{1}$ 
 $\vee (\lambda_{1} \oplus \kappa_{1})((\lambda_{1} \oplus k_{1}) \vee \overline{m_{1}})$ 
 $\sigma_{1} = b_{1}$ 
 $s_{1} = d_{1}$ 

2.  $d_{1} = m_{1} \oplus \lambda_{1} \oplus k_{1} \oplus k_{1}$ 
 $m_{1-1} = \kappa_{1} \vee \overline{\lambda_{1}} \lambda_{1} k_{1} \overline{k_{1}}$ 
 $b_{1-1} = \overline{m_{1}} \overline{\lambda_{1}} k_{1} \vee \overline{m_{1}} \lambda_{1} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{d_{1}} b_{1}$ 

3.  $d_{1} = m_{1} \oplus \lambda_{1} \oplus \lambda_{1} \oplus k_{1}$ 
 $m_{1-1} = \overline{\lambda_{1}} \lambda_{1} k_{1} \vee \overline{m_{1}} \lambda_{1} \overline{k_{1}} \vee \lambda_{1} \lambda_{1} k_{1}$ 
 $\sigma_{1} = b_{1}$ 
 $s_{1} = d_{1} \oplus b_{1}$ 

4.  $d_{1} = m_{1} \oplus \overline{\lambda_{1}} \overline{\lambda_{1}} \oplus \overline{\kappa_{1}} \overline{k_{1}}$ 
 $b_{1-1} = \overline{\kappa_{1}} \overline{k_{1}} (\overline{m_{1}} \vee \overline{\lambda_{1}} \overline{k_{1}}) \vee \overline{m_{1}} \overline{\lambda_{1}} \overline{\lambda_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} b_{1}$ 

5.  $d_{1} = m_{1} \oplus \lambda_{1} \oplus \lambda_{1} \oplus \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} b_{1} \oplus \overline{k_{1}} \oplus \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} b_{1} \oplus \overline{k_{1}} \oplus \overline{k_{1}} \overline{k_{1}} \vee \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} b_{1} \oplus \overline{\lambda_{1}} \overline{k_{1}} \vee \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} b_{1} \oplus \overline{\lambda_{1}} \overline{k_{1}} \vee \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} k_{1}$ 
 $\sigma_{1} = \overline{\lambda_{1}} \vee \overline{\lambda_{1}} \overline{k_{1}} \vee \overline{\mu_{1}} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}} \overline{k_{1}} \vee \lambda_{1} \overline{k_{1}} \overline{k_{1}$ 

Table 4(a) Boolean equations for symmetric subtracters with formats 1 to 6

$$7. d_{i} = m_{i} \bigoplus_{\overline{\lambda}_{i}} \overline{k}_{i} \bigoplus_{\overline{\kappa}_{i}} \overline{k}_{i}$$

$$m_{i-1} = \overline{\lambda}_{i} \overline{k}_{i} \bigvee_{\overline{\lambda}_{i}} \overline{k}_{i}$$

$$b_{i-1} = \overline{m}_{i} \overline{\lambda}_{i} \overline{k}_{i} \bigvee_{\overline{m}_{i}} \overline{k}_{i} \overline{k}_{i}$$

$$\nabla_{i} = b_{i} \overline{d}_{i}$$

$$S_{i} = d_{i} \overline{b}_{i}$$

$$8. d_{i} = m_{i} \bigoplus_{\overline{\lambda}_{i}} \lambda_{i} \overline{k}_{i} \bigoplus_{\overline{k}_{i}} \overline{k}_{i}$$

$$m_{i-1} = \kappa_{i} (\overline{\lambda}_{i} \bigvee_{\overline{\lambda}_{i}} \overline{k}_{i})$$

$$b_{i-1} = m_{i} \lambda_{i} \overline{k}_{i} \bigvee_{\overline{\lambda}_{i}} \lambda_{i} \overline{k}_{i} \overline{k}_{i} \bigvee_{\overline{m}_{i}} \kappa_{i} \overline{k}_{i}$$

$$\nabla_{i} = \overline{d}_{i} \bigvee_{\overline{k}_{i}} b_{i}$$

$$9. d_{i} = m_{i} \bigoplus_{\overline{\lambda}_{i}} \lambda_{i} \bigcup_{\overline{\lambda}_{i}} A_{i} \overline{k}_{i} \bigvee_{\overline{\lambda}_{i}} \overline{k}_{i} \bigvee_{\overline{k}_{i}} \lambda_{i} \overline{k}_{i} \bigvee_{\overline{k}_{i}} K_{i}$$

$$m_{i-1} = \overline{\lambda}_{i} \lambda_{i} \bigvee_{\overline{\lambda}_{i}} A_{i} \overline{k}_{i} \bigvee_{\overline{\lambda}_{i}} \overline{k}_{i} \bigvee_{\overline{k}_{i}} \lambda_{i} \overline{k}_{i} \bigvee_{\overline{k}_{i}} K_{i}$$

$$b_{i-1} = \overline{m}_{i} \kappa_{i} \overline{k}_{i} \bigvee_{\overline{k}_{i}} \overline{k}_{i} \bigvee_{\overline{k}_{i}} \lambda_{i} \lambda_{i} \lambda_{i} \lambda_{i} \lambda_{i} k_{i}$$

$$\nabla_{i} = \overline{d}_{i} \bigvee_{\overline{k}_{i}} b_{i}$$

$$\sigma_{i} = \overline{d}_{i} \bigvee_{\overline{k}_{i}} b_{i}$$

Table 4(b) Boolean equations for symmetric subtracters with formats 7 to 9

#### Appendix

Tables 1 and 2 contain the lower block designs given by Borovec in reference 4. As shown in Figure 2, the inputs to the lower block of the (i) th unit are  $k_1^* = (\lambda_1, l_1)$ ,  $k_i = (\kappa_1, k_1)$  and  $m_{i-1}$ ; the outputs are  $m_i$ ,  $d_{i-1}$  and  $b_{i-2}$ . The output value  $m_i + 2 \cdot d_{i-1} + 4 \cdot b_{i-2}$  is unique in this arrangement.

For the symmetric adder designs the digit sets for  $b_1$ ,  $m_1$ , and  $d_1$  are  $\{0, 1\}$   $\{0, 1\}$  and  $\{\overline{1}, 0\}$  respectively. For the symmetric subtracter, the corresponding sets are  $\{\overline{1}, 0\}$ ,  $\{\overline{1}, 0\}$  and  $\{0, 1\}$ .

$$\begin{vmatrix} 1. & \mathbf{m_i} &= \kappa_i & \oplus \lambda_i & \oplus \mathbf{1_i} & \oplus \kappa_i \\ \mathbf{d_{i-1}} &= & \mathbf{m_{i-1}} & \oplus (\bar{\kappa}_i \bar{k}_i \vee \lambda_i \bar{1}_i \vee \bar{\kappa}_i k_i (\bar{\lambda}_i \vee 1_i) \\ & & \vee \bar{\lambda}_i \mathbf{1_i} (\bar{\kappa}_i \vee k_i)) \\ \mathbf{b_{i-2}} &= & \kappa_i \bar{k}_i \lambda_i \mathbf{1_i} \vee \mathbf{m_{i-1}} (\lambda_i \mathbf{1_i} \vee \kappa_i \bar{k}_i \vee \bar{1}_i \bar{k}_i) \\ & & \vee \kappa_i \bar{\lambda}_i) \end{vmatrix}$$

2. 
$$m_{i} = k_{i} \oplus l_{i}$$

$$d_{i-1} = \kappa_{i} \lambda_{i} \vee \overline{m}_{i-1} \overline{\kappa}_{i} \overline{\lambda}_{i} (k_{i} \vee \overline{l}_{i})$$

$$\vee m_{i-1} (\lambda_{i} \vee \overline{l}_{i} \kappa_{i} \vee \overline{l}_{i} \overline{k}_{i})$$

$$b_{i-2} = \kappa_{i} \lambda_{i} \vee m_{i-1} (\lambda_{i} \vee \kappa_{i} \vee \overline{l}_{i} \overline{k}_{i})$$

3. 
$$\begin{split} \mathbf{m_i} &= \mathbf{k_i} \oplus \mathbf{l_i} \\ \mathbf{d_{i-1}} &= \mathbf{m_{i-1}} \oplus (\overline{\kappa_i} \overline{\lambda_i} \mathbf{l_i} \vee \overline{\mathbf{k_i}} \overline{\lambda_i} \mathbf{l_i} \vee \overline{\kappa_i} \mathbf{k_i} \overline{\mathbf{l_i}} \\ & \vee \kappa_i \mathbf{k_i} \lambda_i \mathbf{l_i}) \\ \mathbf{b_{i-2}} &= \kappa_i \mathbf{k_i} \lambda_i \mathbf{l_i} \vee \mathbf{m_{i-1}} (\kappa_i \mathbf{k_i} \vee \lambda_i \mathbf{l_i} \vee \overline{\mathbf{k_i}} \overline{\mathbf{l_i}}) \end{split}$$

$$\begin{array}{lll} 4. & \mathbf{m_i} &= & \overline{\kappa_i} \overline{k_i} \oplus \overline{\lambda_i} \overline{1_i} \\ & \mathbf{d_{i-1}} &= & \overline{m_{i-1}} (\overline{\kappa_i} \mathbf{1_i} \vee \kappa_i \overline{\lambda_i} \vee \kappa_i \lambda_i) \\ & & \vee \mathbf{m_{i-1}} (\kappa_i \oplus \lambda_i) \\ & \mathbf{b_{i-2}} &= & \kappa_i \lambda_i \vee \mathbf{m_{i-1}} (\kappa_i \vee \lambda_i \vee \overline{k_i} \overline{1_i}) \end{array}$$

5. 
$$m_{i} = k_{i} \oplus l_{i}$$

$$d_{i-1} = \overline{m}_{i-1}(\kappa_{i}k_{i}\lambda_{i}l_{i} \vee \overline{\kappa}_{i}\overline{\lambda}_{i} \vee \overline{\kappa}_{i}\overline{l}_{i} \vee \overline{k}_{i}\overline{\lambda}_{i})$$

$$\vee m_{i-1}(\kappa_{i}\overline{l}_{i} \vee \overline{\kappa}_{i}\lambda_{i}l_{i} \vee \kappa_{i}k_{i}\overline{l}_{i}$$

$$\kappa_{i}k_{i}\overline{\lambda}_{i})$$

$$b_{i-2} = \kappa_{i}k_{i}\lambda_{i}l_{i} \vee m_{i-1}(\kappa_{i}k_{i} \vee \lambda_{i}l_{i} \vee \lambda_{i}\kappa_{i})$$

$$\begin{vmatrix} 6. & \mathbf{m_i} &= \overline{\mathbf{k_i}} \overline{\mathbf{k_i}} & \oplus \overline{\lambda_i} \overline{\mathbf{l_i}} \\ & \mathbf{d_{i-1}} &= \mathbf{m_{i-1}} & \oplus ((\overline{\lambda_i} \vee \overline{\mathbf{l_i}}) (\kappa_i & \oplus k_i) \\ & \vee \overline{\mathbf{k_i}} (\lambda_i & \oplus \mathbf{l_i}) \vee \kappa_i k_i \lambda_i \mathbf{l_i} ) \\ & \mathbf{b_{i-2}} &= \kappa_i k_i \lambda_i \overline{\mathbf{l_i}} \vee \mathbf{m_{i-1}} (\kappa_i k_i \vee \lambda_i \mathbf{l_i} \\ & \vee \overline{\kappa_i} \overline{k_i} \overline{\lambda_i} \overline{\mathbf{l_i}} ) \end{vmatrix}$$

Table 1(a) Lower blocks of symmetric adders with formats 1 to 7

8. 
$$m_{i} = \kappa_{i} \overline{k}_{i} \oplus \lambda_{i} \overline{1}_{i}$$

$$d_{i-1} = m_{i-1} \oplus (\overline{\kappa}_{i} \overline{\lambda}_{i} \vee \overline{k}_{i} \overline{\lambda}_{i} \vee \overline{\kappa}_{i} \overline{1}_{i}$$

$$\vee \kappa_{i} k_{i} \lambda_{i} 1_{i})$$

$$b_{i-2} = \kappa_{i} k_{i} \lambda_{i} 1_{i} \vee m_{i-1} (\kappa_{i} k_{i} \vee \kappa_{i} \lambda_{i} \vee \overline{\kappa}_{i} \lambda_{i} \vee \overline{\kappa}_{i} \lambda_{i} \vee \overline{\kappa}_{i} \lambda_{i} \wedge \overline{\kappa}_{i} \wedge \overline{\kappa}_{i}$$

Table 1(b) Lower blocks of symmetric adders with formats 8 and 9

2. 
$$m_{1} = k_{1} \oplus l_{1}$$

$$d_{1-1} = m_{1-1} \oplus (\kappa_{1}\lambda_{1} \vee \overline{k}_{1}\overline{l}_{1} \vee \overline{k}_{1}\lambda_{1})$$

$$b_{1-2} = \overline{\kappa_{1}}k_{1}\overline{\lambda_{1}}l_{1} \vee m_{1-1}(\overline{\lambda_{1}}l_{1} \vee \overline{\kappa_{1}}k_{1} \vee \overline{\kappa_{1}}\overline{\lambda_{1}})$$

3. 
$$m_{i} = k_{i} \oplus l_{i}$$

$$d_{i-1} = m_{i-1} \oplus (\kappa_{i}\lambda_{i}l_{i} \vee \overline{k}_{i}\lambda_{i}l_{i} \vee \kappa_{i}k_{i}\lambda_{i}$$

$$\vee \overline{\kappa}_{i}k_{i}\overline{\lambda}_{i}l_{i})$$

$$b_{i-2} = \overline{\kappa}_{i}k_{i}\overline{\lambda}_{i}l_{i} \vee \overline{m}_{i-1}\kappa_{i}k_{i}\lambda_{i}l_{i}$$

$$\vee m_{i-1}(\overline{\kappa}_{i}k_{i} \vee \overline{\lambda}_{i}l_{i} \vee k_{i}l_{i})$$

4. 
$$\begin{array}{lll} \texttt{m_i} &=& \overline{\kappa_i} \overline{k_i} & \oplus & \overline{\lambda_i} \overline{l_i} \\ \texttt{d_{i-1}} &=& \texttt{m_{i-1}} & \oplus & (k_i l_i \vee \kappa_i \overline{l_i} \vee \overline{k_i} \lambda_i) \\ \texttt{b_{i-2}} &=& k_i l_i \vee \texttt{m_{i-1}} (k_i \vee l_i \vee \overline{\kappa_i} \overline{\lambda_i}) \end{array}$$

5. 
$$m_i = k_i \oplus l_i$$

$$d_{i-1} = \overline{m}_{i-1} \oplus ((\kappa_i \oplus \lambda_i) \vee \overline{k}_i \overline{l}_i)$$

$$b_{i-2} = \overline{\kappa}_i \overline{\lambda}_i \vee m_{i-1} (\overline{\kappa}_i \vee \overline{\lambda}_i \vee \overline{k}_i \overline{l}_i)$$

Table 2(a) Lower blocks of symmetric subtracters with formats 1 to 5

Table 2(b) Lower blocks of symmetric subtracters with formats 6 to 9

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